

# Revisiting the calculation of effective free distance of turbo codes

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The expression for the minimum Hamming weight of the output of a constituent convolutional encoder, when its input is a weight-2 sequence is revisited. The new expression particularly facilitates the calculation of the effective free distance of recently proposed schemes, namely non-systematic turbo codes and pseudo-randomly punctured turbo codes.

**Introduction:** Several authors [1, 2] have agreed that the performance of turbo codes [3] at the error floor region is largely determined by the weight-2 input minimum distance, which corresponds to the minimum Hamming weight among all codeword sequences generated by input sequences of weight two. If a turbo code  $T$  consists of  $N$  parallel concatenated convolutional codes separated by uniform interleavers, its weight-2 input minimum distance  $d_2^T$ , which is also referred to as the *effective free distance* of  $T$ , can be written as [4, 5]

$$d_2^T = \sum_{k=1}^N d_2^{(k)} \quad (1)$$

where  $d_2^{(k)}$  is the weight-2 input minimum distance of the  $k$ th constituent code.

Bounds on the weight-2 input minimum distance  $d_2$  of a convolutional code as well as exact expressions are provided in [1, 4, 6]. Nevertheless, the exact expressions are accurate only when either the impulse response of the code is known [6] or the structure of the code meets particular criteria [1, 4]. Recently, Banerjee *et al.* [5] demonstrated that non-systematic turbo codes using quick-look-in (QLI) convolutional codes as constituent codes, can achieve lower error floors than those of conventional systematic turbo codes. Unfortunately QLI codes do not always meet the conditions of [1, 4], hence the corresponding expressions cannot be used to determine their weight-2 input minimum distances. In this Letter we relax the conditions of [1, 4] and present expressions which allow the accurate calculation of  $d_2$  for a wider set of convolutional codes.

**Preliminaries:** Let  $(r, 1, v)$  represent a rate-1/ $r$  convolutional code of memory  $v$  and  $\mathbf{G}(D) = [\mathbf{P}^{(1)}(D)/\mathbf{Q}(D), \dots, \mathbf{P}^{(r)}(D)/\mathbf{Q}(D)]$  be the generator matrix of the recursive encoder for that code, where  $\mathbf{P}^{(i)}(D) = p_v^{(i)} D^v + \dots + p_1^{(i)} D + p_0^{(i)}$  denotes the  $i$ th feed-forward generator polynomial and  $\mathbf{Q}(D) = q_v D^v + \dots + q_1 D + q_0$  corresponds to the feedback generator polynomial, with coefficients  $p_j^{(i)}, q_j \in \{0, 1\}$ . Note that none of the feed-forward polynomials is equal to  $\mathbf{Q}(D)$ , while  $\mathbf{P}^{(1)}(D)/\mathbf{Q}(D) = 1$  only if the convolutional code is systematic.

It was shown in [1, 4] that the weight-2 input minimum distance of a  $(r, 1, v)$  recursive convolutional code is given by  $d_2 = r(2 + 2^{v-1})$  if the code is non-systematic and  $d_2 = 2 + (r-1)(2 + 2^{v-1})$  if the code is systematic. In both cases, it has been assumed that  $\mathbf{Q}(D)$  is a primitive polynomial of order  $v \geq 2$ , i.e.  $\deg \mathbf{Q}(D) = v$ , while  $\mathbf{P}^{(i)}(D)$  is a monic polynomial with constant term 1, i.e.  $p_v^{(i)} = p_0^{(i)} = 1$ . Consequently,  $\deg \mathbf{P}^{(i)}(D) = \deg \mathbf{Q}(D) = v$ .

**Calculation of  $d_2$  when  $\deg \mathbf{P}^{(i)}(D) \leq \deg \mathbf{Q}(D)$ :** As previously, we assume that  $\mathbf{Q}(D)$  is a primitive polynomial of order  $v \geq 2$ , since it has been shown that turbo codes using primitive feedback generator polynomials yield an excellent performance [1]. Let  $u(t)$  denote the input bit to the encoder at time step  $t$  and  $r_m(t)$  represent the output of the  $m$ th memory element, where  $m = 1, \dots, v$ . Initially, we focus on the  $i$ th non-systematic output of the encoder. The corresponding output bit  $y^{(i)}(t)$  can be expressed as follows

$$y^{(i)}(t) = p_0^{(i)} u(t) \oplus (p_1^{(i)} \oplus q_1 p_0^{(i)}) r_1(t) \oplus \dots \oplus (p_{v-1}^{(i)} \oplus q_{v-1} p_0^{(i)}) r_{v-1}(t) \oplus (p_v^{(i)} \oplus p_0^{(i)}) r_v(t) \quad (2)$$

where the symbol  $\oplus$  denotes the mod-2 addition. We have also adopted the notation  $d_{t_1 \rightarrow t_2}^{(i)}$  to represent the weight of the sequence generated by the  $i$ th non-systematic output of the encoder during the transition from

time step  $t_1$  to time step  $t_2$ , i.e.

$$d_{t_1 \rightarrow t_2}^{(i)} = \sum_{t=t_1}^{t_2-1} y^{(i)}(t) \quad (3)$$

If  $L$  is the period of the primitive feedback polynomial  $\mathbf{Q}(D)$ , the two nonzero bits of a weight-2 input sequence should be separated by  $L-1$  zeroes such that the encoder returns to the zero state [7], i.e.  $r_m(t) = 0$  for all  $m$ . Let  $u(0) = u(L) = 1$ , while  $u(1) = \dots = u(L-1) = 0$ . Note that the weight-2 input minimum distance of the  $i$ th non-systematic output of the encoder is quantified by  $d_{0 \rightarrow L+1}^{(i)}$ . For convenience, we express  $d_{0 \rightarrow L+1}^{(i)}$  as  $d_{0 \rightarrow L+1}^{(i)} = d_{0 \rightarrow 1}^{(i)} + d_{1 \rightarrow L}^{(i)} + d_{L \rightarrow L+1}^{(i)}$  and compute each term separately:

- $t: 0 \rightarrow 1$  – Assuming that the encoder was initialised to the zero state, we obtain  $d_{0 \rightarrow 1}^{(i)} = y^{(i)}(0) = p_0^{(i)}$  from (2) and (3), since  $u(0) = 1$  and  $r_1(0) = \dots = r_v(0) = 0$ .
- $t: 1 \rightarrow L$  – Let us first consider the case when  $t: 1 \rightarrow L+1$  and  $u(L) = 0$ . Owing to the properties of primitive polynomials, the output stream is a pseudo-noise sequence having weight  $d_{1 \rightarrow L+1}^{(i)} = 2^{v-1}$ , given that  $\mathbf{P}^{(i)}(D) \neq \mathbf{Q}(D)$  [7]. Furthermore, when  $t = L$ , the encoder is in state 1 [7], i.e.  $r_1(L) = \dots = r_{v-1}(L) = 0$  and  $r_v(L) = 1$ . Hence, if  $u(L) = 0$  is the input bit, the encoder outputs  $y^{(i)}(L) = p_v^{(i)} \oplus p_0^{(i)}$ , which is also the value of  $d_{L \rightarrow L+1}^{(i)}$ . However, an equivalent and more convenient form of the previous expression for the output weight is  $d_{L \rightarrow L+1}^{(i)} = (p_v^{(i)} - p_0^{(i)})^2$ . Consequently, we can compute the target quantity  $d_{1 \rightarrow L}^{(i)}$  by subtracting  $d_{L \rightarrow L+1}^{(i)}$  from  $d_{1 \rightarrow L+1}^{(i)}$  and obtain  $d_{1 \rightarrow L}^{(i)} = 2^{v-1} - (p_v^{(i)} - p_0^{(i)})^2$ , independently of the value of  $u(L)$ .
- $t: L \rightarrow L+1$  – We established that if  $t = L$  then  $r_v(L) = 1$ , while the output of the remaining memory elements is zero. That is when the second nonzero bit, namely  $u(L) = 1$ , of the weight-2 sequence is input to the encoder the bit forces the encoder to return to the zero state. Using (2) and (3), we find that  $d_{L \rightarrow L+1}^{(i)} = y^{(i)}(L) = p_v^{(i)}$ .

Thus, the weight of the  $i$ th non-systematic output sequence of the encoder for a weight-2 input sequence can be expressed as

$$\begin{aligned} d_{0 \rightarrow L+1}^{(i)} &= d_{0 \rightarrow 1}^{(i)} + d_{1 \rightarrow L}^{(i)} + d_{L \rightarrow L+1}^{(i)} \\ &= p_0^{(i)} + 2^{v-1} - (p_v^{(i)} - 2p_0^{(i)} p_v^{(i)} + p_0^{(i)}) + p_v^{(i)} \\ &= 2^{v-1} + 2p_0^{(i)} p_v^{(i)}, \end{aligned} \quad (4)$$

using the fact that the value of a binary number, such as  $p_j^{(i)}$ , does not alter when it is raised to a power (e.g.,  $(p_j^{(i)})^2 = p_j^{(i)}$ ). The overall weight-2 input minimum distance of the rate-1/ $r$  recursive convolutional encoder can be obtained as follows

$$\begin{aligned} d_2 &= \sum_{i=1}^r d_{0 \rightarrow L+1}^{(i)} \\ &= \begin{cases} r2^{v-1} + 2 \sum_{i=1}^r p_0^{(i)} p_v^{(i)}, & \text{if the code is non-systematic} \\ 2 + (r-1)2^{v-1} + 2 \sum_{i=2}^r p_0^{(i)} p_v^{(i)}, & \text{if the code is systematic,} \end{cases} \end{aligned} \quad (5)$$

**Extension to pseudorandomly punctured codes:** Pseudorandom (PR) puncturing, initially introduced in [7], is a method to increase the rate of a constituent recursive systematic convolutional code with generator matrix  $\mathbf{G}(D) = [\mathbf{P}(D)/\mathbf{Q}(D)]$  from 1/2 to 1 by periodically eliminating particular bits from its output. Note that  $\mathbf{Q}(D)$  should be primitive. It has been shown [8] that a rate-1/2 turbo code consisting of a rate-1 PR-punctured convolutional code and a rate-1 non-systematic convolutional code, yields a lower error floor than that of its rate-1/3 parent code. Following a similar reasoning as in the precoding section, we can express (the proof has been omitted) the weight-2 input minimum distance of a PR-punctured convolutional code  $(1, 1, v)$  as

$$d_2 = 2^{v-2} + 2p_0 p_v. \quad (6)$$

**Conclusion:** In this Letter we expressed the weight-2 input minimum distance of a rate-1/ $r$  convolutional code as a function of the coefficients of its feed-forward generator polynomials  $\mathbf{P}^{(i)}(D)$ , with  $i = 1, \dots, r$ , for a primitive feedback generator polynomial  $\mathbf{Q}(D)$ . This expression can be

used to accurately compute the effective free distance of both conventional systematic turbo codes as well as non-systematic turbo codes [5, 8] that consist of convolutional codes with  $\deg \mathbf{P}^{(i)}(D) \leq \deg \mathbf{Q}(D)$ .

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## References

- 1 Benedetto, S., and Montorsi, G.: 'Design of parallel concatenated convolutional codes', *IEEE Trans. Commun.*, 1996, **COM-44**, pp. 591–600
- 2 Perez, L.C., Seghers, J., and Costello, D.J. Jr.: 'A distance spectrum interpretation of turbo codes', *IEEE Trans. Inf. Theory*, 1996, **IT-42**, pp. 1698–1709
- 3 Berrou, C., Glavieux, A., and Thitimajshima, P.: 'Near Shannon limit error-correcting coding and decoding: turbo-codes'. Proc. ICC, Geneva, Switzerland, May 1993
- 4 Divsalar, D., and McEliece, R.J.: 'The effective free distance of turbo codes', *Electron. Lett.*, 1996, **32**, (5), pp. 445–446
- 5 Banerjee, A., Vatta, F., Scanavino, B., and Costello, D.J. Jr.: 'Nonsystematic turbo codes', *IEEE Trans. Commun.*, 2005, **COM-53**, pp. 1841–1849
- 6 Yoscovich, I., and Snyders, J.: 'On the effective free distance of turbo codes'. Proc. ITW, Killarney, Ireland, June 1998
- 7 Chatzigeorgiou, I., Rodrigues, M.R.D., Wassell, I.J., and Carrasco, R.: 'A union bound approximation for rapid performance evaluation of punctured turbo codes'. Proc. CISS, Baltimore, MD, USA, March 2007
- 8 Chatzigeorgiou, I., Rodrigues, M.R.D., Wassell, I.J., and Carrasco, R.: 'Pseudo-random puncturing: a technique to lower the error floor of turbo codes'. Proc. ISIT, Nice, France, June 2007